

**STOCHASTIC PROCESSES. SOLUTIONS TO HOME
ASSIGNMENTS**

2. STATIONARY RANDOM PROCESSES

Problem 2.1

It is well known, that characteristic function of a random vector defines its distribution. Introduce the vector

$$\xi_n = \begin{bmatrix} \xi(t_1) \\ \xi(t_2) \\ \dots \\ \xi(t_n) \end{bmatrix}$$

In this case

$$\Phi(\lambda) \triangleq \mathbb{E}e^{i\lambda^T \xi_n} = \mathbb{E}\mathbb{E}(e^{i\lambda^T \xi_n} | \alpha, \beta)$$

since γ is independent of α, β we get

$$\Phi(\lambda) = \mathbb{E} \frac{1}{2\pi} \int_0^{2\pi} \exp \left(\sum_{k=1}^n i \lambda_k \alpha \sin(\beta t_k + \gamma) \right) d\gamma$$

Denote by $\Phi_h(\lambda)$ the characteristic function of the time shifted vector, namely

$$\begin{aligned} \Phi_h(\lambda) &= \mathbb{E}\mathbb{E} \left[\exp \left\{ i \sum_{k=1}^n \lambda_k \xi(t_k + h) \right\} \middle| \alpha, \beta \right] = \\ &= \mathbb{E} \frac{1}{2\pi} \int_0^{2\pi} \exp \left\{ i \sum_{k=1}^n \lambda_k \alpha \sin(\beta t_k + \beta h + \gamma) \right\} d\gamma = \\ &= \mathbb{E} \frac{1}{2\pi} \int_{\beta h}^{2\pi + \beta h} \exp \left\{ i \sum_{k=1}^n \lambda_k \alpha \sin(\beta t_k + \gamma') \right\} d\gamma' = \\ &= \mathbb{E} \frac{1}{2\pi} \int_{\beta h}^{2\pi} \exp \left\{ i \sum_{k=1}^n \lambda_k \alpha \sin(\beta t_k + \gamma') \right\} d\gamma' + \\ &\quad + \mathbb{E} \frac{1}{2\pi} \int_{2\pi}^{2\pi + \beta h} \exp \left\{ i \sum_{k=1}^n \lambda_k \alpha \sin(\beta t_k + \gamma') \right\} d\gamma' = \\ &= \mathbb{E} \frac{1}{2\pi} \int_0^{2\pi} \exp \left\{ i \sum_{k=1}^n \lambda_k \alpha \sin(\beta t_k + \gamma'') \right\} d\gamma'' \equiv \Phi(\lambda) \end{aligned}$$

Problem 2.2

(a) $R(k)$ is non negative definite if

$$\sum_{k,m} a_k R(k-m) \bar{a}_m \geq 0 \quad (2.1)$$

for any sequence $\{a_k\}$. Let $S(\lambda)$ be spectral density corresponding to $R(k)$, then

$$R(k-m) = \frac{1}{2\pi} \int_{[-\pi, \pi]} S(\lambda) e^{j(k-m)\lambda} d\lambda$$

and

$$\begin{aligned} \sum_{k,m} a_k R(k-m) \bar{a}_m &= \sum_{k,m} a_k \frac{1}{2\pi} \int_{[-\pi, \pi]} S(\lambda) e^{j(k-m)\lambda} d\lambda \bar{a}_m = \\ &= \frac{1}{2\pi} \int_{[-\pi, \pi]} S(\lambda) \sum_{k,m} a_k e^{jk\lambda} \bar{a}_m e^{-jm\lambda} d\lambda = \\ &= \frac{1}{2\pi} \int_{[-\pi, \pi]} S(\lambda) |A(\lambda)|^2 d\lambda \end{aligned} \quad (2.2)$$

where $A(\lambda)$ is the Fourier transform of \bar{a}_k . Due to (2.1), (2.2) and arbitrariness of a_k , $S(\lambda) \geq 0$ follows. Starting from $S(\lambda) \geq 0$, by (2.2), we deduce (2.1), which proves the other direction.

(b) Assume that $R(n)$ can be decomposed

$$R(n) = \sum_{k=-\infty}^{\infty} h(k) \bar{h}(k-n)$$

Then

$$\begin{aligned} S(\lambda) &= \sum_m R(m) e^{-j\lambda m} = \sum_m \sum_k h(k) \bar{h}(k-m) e^{-j\lambda m} = \\ &= \sum_k h(k) \sum_{\ell} \bar{h}(\ell) e^{-j\lambda(k-\ell)} = |H(\lambda)|^2 \geq 0 \end{aligned}$$

for any λ . So, by virtue of (a), $R(n)$ is a non negative definite sequence.

(c) Let X'_n and X''_n be a pair of independent processes with zero mean and correlation functions $R'(k, m)$ and $R''(k, m)$. Introduce $Y_n = X'_n X''_n$ and $Z_n = X'_n + X''_n$. Then

$$\mathbb{E}Y_k Y_m = \mathbb{E}X'_k X''_k X'_m X''_m = \mathbb{E}X'_k X'_m \mathbb{E}X''_k X''_m = R'(k, m) R''(k, m)$$

and

$$\mathbb{E}Z_k Z_m = \mathbb{E}(X'_k + X''_k)(X'_m + X''_m) = R'(k, m) + R''(k, m)$$

Problem 2.3

Any symmetric sequence $R(n)$, which satisfies¹

- (i) $R(0) \geq R(m)$, $m \neq 0$
- (ii) $R(n)$ is positive definite

¹Note that these conditions are not necessarily independent

can be an autocorrelation function of some process.

- (a) For $R(n) = e^{-n^2}$ (i) is obvious. Verify (ii) using the results of the previous problem

$$\begin{aligned} S(\lambda) &= \sum_n R(n)e^{-jn\lambda} = \sum_n e^{-n^2-jn\lambda} = 1 + 2 \sum_{n=1}^{\infty} e^{-n^2} \cos(n\lambda) \geq \\ &\geq 1 - 2 \sum_{n=1}^{\infty} e^{-n^2} \geq 1 - 2e^{-1} - 2e^{-4} - 2 \sum_{n=3}^{\infty} e^{-n} = \\ &= 1 - 2e^{-1} - 2e^{-4} - 2e^{-3}/(1 - e^{-1}) > 0, \quad \forall \lambda \end{aligned}$$

so that (ii) holds as well.

- (b) No: $S(\lambda) = 1 + 1.4 \cos(\lambda)$ is negative for λ on some interval (e.g. around $\lambda = \pi$)
(c) Note that $R(n) = h(n) \star h(-n)$ where $h(n) = I(0 \leq n < N)$, so by virtue of (b) from the previous problem, $R(n)$ is non negative definite.

Problem 2.4

- (a) Note that

$$\lambda_k = \frac{v_k^* R_x v_k}{v_k^* v_k} \quad (2.3)$$

where v_k is the eigenvector corresponding to λ_k . Denote by $v_{k,\ell}$ the ℓ -th component of the k -th eigenvector. Then

$$v_k^* R_x v_k = \sum_{\ell=1}^N \sum_{m=1}^N v_{k,\ell} R_x(\ell, m) v_{k,m} = \sum_{\ell=1}^N \sum_{m=1}^N v_{k,\ell} r_x(\ell - m) v_{k,m}$$

where $r_x(\ell - m) = \mathbb{E}X(\ell)X(m)$ is the autocorrelation sequence of the process. Using the representation

$$r_x(\ell - m) = \frac{1}{2\pi} \int_{[-\pi, \pi]} S_x(\lambda) e^{j\lambda(\ell - m)} d\lambda$$

obtain

$$\begin{aligned} v_k^* R_x v_k &= \frac{1}{2\pi} \int_{[-\pi, \pi]} S_x(\lambda) \left\{ \sum_{\ell} v_{k,\ell} e^{j\lambda\ell} \sum_m v_{k,m} e^{-j\lambda m} \right\} d\lambda = \\ &= \frac{1}{2\pi} \int_{[-\pi, \pi]} S_x(\lambda) |V_k(\lambda)|^2 d\lambda \end{aligned}$$

Similarly

$$v_k^* v_k = \frac{1}{2\pi} \int_{[-\pi, \pi]} |V_k(\lambda)|^2 d\lambda$$

so that by (2.3)

$$\lambda_k = \frac{\int_{[-\pi, \pi]} S_x(\lambda) |V_k(\lambda)|^2 d\lambda}{\int_{[-\pi, \pi]} |V_k(\lambda)|^2 d\lambda}$$

which in turn implies

$$\min_{\lambda} S_x(\lambda) \leq \lambda_k \leq \max_{\lambda} S_x(\lambda)$$

for all k .

(b) Introduce

$$\gamma_n = \frac{\mathbb{E}(\sum_{k=0}^{N-1} X_{n-k} a_k)^2}{\mathbb{E}(\sum_{k=0}^{N-1} \xi_{n-k} a_k)^2}$$

Define vectors $X^n = [X_n, \dots, X_{n-N+1}]^*$, $\xi^n = [\xi_n, \dots, \xi_{n-N+1}]^*$ and $a = [a_0, \dots, a_{N-1}]^*$ so that

$$\gamma_n \equiv \gamma = \frac{\mathbb{E}(X^{n*} a)^2}{\mathbb{E}(\xi^{n*} a)^2} = \frac{a^* R_x a}{\sigma^2 a^* a} = \sigma^{-2} \frac{a^* U \Lambda U^* a}{a^* U U^* a}$$

where U is an orthogonal matrix with v_k as columns and Λ is a diagonal matrix with $\Lambda_{jj} = \lambda_j$. Set $\tilde{a} = U^* a$, then

$$\gamma = \sigma^{-2} \frac{\tilde{a}^* \Lambda \tilde{a}}{\tilde{a}^* \tilde{a}} = \sigma^{-2} \frac{\sum_{j=0}^{N-1} \tilde{a}_j^2 \lambda_j}{\sum_{j=0}^{N-1} \tilde{a}_j^2} \leq \lambda_{\max} / \sigma^2$$

where the equality holds when $a = v_{\max}$, the eigenvector corresponding to $\lambda_{\max} = \max_j \lambda_j$.

3. LINEAR ESTIMATION OF STATIONARY SEQUENCES

Problem 3.1

(a)

$$\widehat{X}_n = \sum_{k=-\infty}^{\infty} Y_k \tilde{a}_{n-k} = \sum_{k=-\infty}^{\infty} Y_{n-k} \tilde{a}_k$$

By orthogonality principle

$$\mathbb{E}(X_n - \widehat{X}_n)Y_{n-\ell} = 0, \quad \ell = \dots, -1, 0, 1, \dots$$

which implies:

$$R_{xy}(\ell) - \sum_k R_y(\ell - k) \tilde{a}_k = 0, \quad \ell = \dots, -1, 0, 1, \dots$$

This version of Wiener-Hopf equation can be solved in the domain of Fourier transform:

$$\begin{aligned} S_{xy}(\lambda) &:= \sum_{\ell} R_{xy}(\ell) e^{-j\lambda\ell} = \sum_k \sum_{\ell} R_y(\ell - k) \tilde{a}_k e^{-j\lambda\ell} = \\ &= \sum_k \tilde{a}_k e^{-j\lambda k} \sum_{\ell} R_y(\ell) e^{-j\lambda\ell} = \tilde{A}(\lambda) S_y(\lambda) \end{aligned}$$

Assuming that $S_y(\lambda) > 0$, we obtain the expression for the filter in terms of spectral densities

$$\tilde{A}(\lambda) = \frac{S_{xy}(\lambda)}{S_y(\lambda)}$$

The mean square error is:

$$\begin{aligned} \mathbb{E}(X_n - \widehat{X}_n)^2 &= \mathbb{E}X_n^2 - \mathbb{E}X_n \widehat{X}_n = \\ &= R_x(0) - \sum_k \mathbb{E}X_n Y_{n-k} \tilde{a}_k = R_x(0) - \sum_k R_{xy}(k) \tilde{a}_k = \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} S_x(\lambda) d\lambda - \frac{1}{2\pi} \int_{-\pi}^{\pi} S_{xy}(\lambda) \sum_k \tilde{a}_k e^{j\lambda k} d\lambda = \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \left(S_x(\lambda) - S_{xy}(\lambda) \tilde{A}(\lambda) \right) d\lambda = \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \left(S_x(\lambda) - \frac{|S_{xy}(\lambda)|^2}{S_y(\lambda)} \right) d\lambda \end{aligned}$$

(b) By orthogonality property:

$$\mathbb{E}(X_n - \sum_{k=0}^{\infty} Y_{n-k} \tilde{a}_k) Y_{\ell} = 0, \quad \ell \leq n$$

and

$$R_{xy}(n - \ell) - \sum_{k=0}^{\infty} R_y(n - \ell - k) \tilde{a}_k = 0, \quad \ell \leq n$$

or

$$R_{xy}(m) - \sum_{k=0}^{\infty} R_y(m - k) \tilde{a}_k = 0, \quad m \geq 0 \tag{3.1}$$

Z -transform of the left hand side of (3.1) reads:

$$S_{xy}(z) - S_y(z)\tilde{A}(z)$$

but only non-positive powers of z of the latter expression obey (3.1), namely:

$$\left[S_{xy}(z) - S_y(z)\tilde{A}(z) \right]_+ = 0$$

where $[\psi(z)]_+$ denotes non-positive powers of the series expansion of $\psi(z)$. Since $S_y(z)$ can be factored:

$$\left[S_{xy}(z) - \tilde{A}(z)B(z)B(1/z) \right]_+ = 0$$

where, say, $B(z)$ is the transform of casual sequence (i.e. its Z transform has only non-positive powers).

$$\left[B(1/z) \left(\frac{S_{xy}(z)}{B(1/z)} - B(z)\tilde{A}(z) \right) \right]_+ = 0$$

Since $B(1/z)$ is the transform of anti-casual sequence, the only way this equation can be satisfied is when $S_{xy}(z)/B(1/z) - \tilde{A}(z)B(z)$ is the transform of anti-casual sequence as well, by other words:

$$\left[S_{xy}(z)/B(1/z) - B(z)\tilde{A}(z) \right]_+ = 0$$

But $\tilde{A}(z)B(z)$ corresponds to a casual sequence, that is

$$\left[\tilde{A}(z)B(z) \right]_+ = \tilde{A}(z)B(z),$$

so the response of the optimal casual filter can be calculated from:

$$\tilde{A}(z) = \frac{1}{B(z)} \left[\frac{S_{xy}(z)}{B(z^{-1})} \right]_+ \quad (3.2)$$

The mean square error can be calculated as in the previous case.

(c) Again orthogonality implies

$$R_{xy}(m) - \sum_{k=0}^p R_y(m-k)\tilde{a}_k = 0, \quad 0 \leq m \leq p \quad (3.3)$$

Define the vectors:

$$\rho_{xy} = \begin{bmatrix} R_{xy}(0) \\ R_{xy}(1) \\ \vdots \\ R_{xy}(p) \end{bmatrix} \quad \tilde{a} = \begin{bmatrix} \tilde{a}_0 \\ \tilde{a}_1 \\ \vdots \\ \tilde{a}_p \end{bmatrix}$$

and the correlation matrix R^y , so that:

$$R^y(i, j) = R_y(i-j), \quad 0 \leq i, j \leq p$$

Now (3.3) has the vector formulation:

$$R^y \tilde{a} = \rho_{xy}$$

and assuming $R^y > 0$, one can obtain the optimal filter:

$$\tilde{a} = [R^y]^{-1} \rho_{xy}$$

The mean square error can be also calculated using these vector notations. Let Y^n denote the vector of $(p+1)$ last samples of Y_n , i.e.

$$Y^n = \begin{bmatrix} Y_n \\ Y_{n-1} \\ \vdots \\ Y_{n-p} \end{bmatrix}$$

$$\begin{aligned} \mathbb{E}(X_n - a^* Y^n)^2 &= R_x(0) - \rho_{xy}^* \tilde{a} - \tilde{a}^* \rho_{xy} + \tilde{a}^* R^y \tilde{a} = \\ &= R_x(0) - \rho_{xy}^* [R^y]^{-1} \rho_{xy} \end{aligned}$$

Problem 3.2

(a) Consider the sequence $(X_n)_{n \in \mathbb{Z}}$, given by:

$$X_n = \sum_{k=-\infty}^n a^{n-k} \varepsilon_k.$$

These series are convergent (for any fixed n , in \mathbb{L}^2) since

$$\xi_m^{(n)} = \sum_{k=-m}^n a^{n-k} \varepsilon_k$$

is a Cauchy sequence and \mathbb{L}^2 is a complete space. Indeed (for, say, $m \geq \ell$)

$$\begin{aligned} \mathbb{E}(\xi_m^{(n)} - \xi_\ell^{(n)})^2 &= \mathbb{E}\left(\sum_{k=-m}^{-\ell} a^{n-k} \varepsilon_k\right)^2 = \sum_{k=-m}^{-\ell} a^{2(n-k)} \leq \sum_{k=-\infty}^{-\ell} a^{2(n-k)} = \\ &= a^{2n} \sum_{k=\ell}^{\infty} a^{2k} = a^{2(n+\ell)} / (1 - a^2) \xrightarrow{\ell \rightarrow \infty} 0. \end{aligned}$$

Clearly X satisfies $X_n = aX_{n-1} + \varepsilon_n$, $n \in \mathbb{Z}$ and it is stationary. Indeed $\mathbb{E}X_n = 0$ for all n and

$$R_x(0) = \mathbb{E}X_n^2 = \mathbb{E}\left(\sum_{k=-\infty}^n a^{n-k} \varepsilon_k\right)^2 = \sum_{k=-\infty}^n a^{2(n-k)} = \sum_{k=0}^{\infty} a^{2k} = \frac{1}{1 - a^2}, \quad \forall n.$$

Then

$$\mathbb{E}X_n X_{n+1} = \mathbb{E}X_n (aX_n + \varepsilon_{n+1}) = aR_x(0)$$

and by induction $\mathbb{E}X_n X_{n+m} = a^{|m|} R_x(0)$, that is the covariance function depends only on the time shift.

- (b) The pair (X, Y) is stationary as well. Clearly $\mathbb{E}Y_n = \mathbb{E}X_n = 0$, and $R_y(k) := \mathbb{E}Y_n Y_{n+k} = R_x(k) + \sigma^2 \delta(k)$ and $R_{xy}(k) := \mathbb{E}X_n Y_{n+k} = R_x(k)$.
- (c) Find the spectral density of X

$$S_x(\lambda) = \sum_{\ell=-\infty}^{\infty} \frac{a^{|\ell|}}{1 - a^2} e^{-j\lambda\ell} = \dots = \frac{1}{1 - 2a \cos \lambda + a^2}$$

and

$$S_y(\lambda) = S_x(\lambda) + 1, \quad S_{xy}(\lambda) = S_x(\lambda)$$

and using the formulas from the previous problem we obtain:

$$A(\lambda) = \frac{S_x}{S_x + 1} = \frac{1}{1 + 1 - 2a \cos \lambda + a^2} = \frac{1}{2 - 2a \cos \lambda + a^2}$$

The minimal mean square error is readily calculated:

$$\begin{aligned} \mathbb{E}(X_n - \hat{X}_n)^2 &= R_x(0) - \sum_k R_{xy}(k)a_k = \frac{1}{2\pi} \int \left[S_x(\lambda) - \frac{|S_{xy}(\lambda)|^2}{S_y(\lambda)} \right] d\lambda \\ &= \frac{1}{2\pi} \int \frac{S_x(\lambda)}{S_x(\lambda) + 1} d\lambda = \frac{1}{2\pi} \int \frac{1}{2 - 2a \cos \lambda + a^2} d\lambda = \frac{1}{\sqrt{4 + a^4}} \end{aligned}$$

(d) Using the formula from the previous problem:

$$\tilde{A}(z) = \frac{1}{B(z)} \left[\frac{S_{xy}(z)}{B(z^{-1})} \right]_+ \quad (3.4)$$

where $B(z)$ is the casual term in the factorization of

$$S_y(z) = B(z)B(z^{-1})$$

In this case

$$S_y(z) = S_x(z) + 1 = \frac{1}{(1 - az)(1 - az^{-1})} + 1 = \frac{a(1 - \gamma z^{-1})(1 - \gamma z)}{\gamma(1 - az^{-1})(1 - az)}$$

where

$$\gamma := \frac{2 + a^2 - \sqrt{4 + a^4}}{2a}$$

Note that $|\gamma| < 1$ for $|a| < 1$. So $B(z)$ is identified as:

$$B(z) := \sqrt{\frac{a}{\gamma}} \frac{1 - \gamma z^{-1}}{1 - az^{-1}}$$

Substitute this into (3.4):

$$\begin{aligned} \tilde{A}(z) &= \sqrt{\frac{\gamma}{a}} \frac{1 - az^{-1}}{1 - \gamma z^{-1}} \left[\frac{\sqrt{\gamma/a}(1 - az)/(1 - \gamma z)}{(1 - az)(1 - az^{-1})} \right]_+ = \\ &= \sqrt{\frac{\gamma}{a}} \frac{1 - az^{-1}}{1 - \gamma z^{-1}} \left[\frac{\sqrt{\gamma/a}}{1 - a\gamma} \left(\frac{1}{1 - az^{-1}} - \frac{1}{1 - \gamma^{-1}z^{-1}} \right) \right]_+ = \\ &= \frac{\gamma}{a} \frac{1}{1 - a\gamma} \frac{1 - az^{-1}}{1 - \gamma z^{-1}} \frac{1}{1 - az^{-1}} = \frac{\gamma}{a(1 - a\gamma)} \frac{1}{1 - \gamma z^{-1}} = \\ &= \frac{2 + a^2 - \sqrt{4 + a^4}}{a^2(\sqrt{4 + a^4} - a^2)} \frac{1}{1 - \gamma z^{-1}} = \frac{a^2 - 2 + \sqrt{4 + a^4}}{2a^2} \frac{1}{1 - \gamma z^{-1}} \end{aligned}$$

The filtering error can be calculated directly, using the formulas similar to the previous case. It also equals the steady state error of the Kalman filter (why?).

(e) Recall that

$$R_y(m) = R_x(m) + 1 \cdot \delta(m), \quad R_{xy}(m) = R_x(m)$$

hence (\tilde{a} now denotes a 2-by-1 vector)

$$\begin{aligned}\tilde{a} &= \begin{pmatrix} R_x(0) + 1 & R_x(1) \\ R_x(1) & R_x(0) + 1 \end{pmatrix}^{-1} \begin{pmatrix} R_x(0) \\ R_x(1) \end{pmatrix} = \\ &= \begin{pmatrix} \frac{1}{1-a^2} + 1 & \frac{a}{1-a^2} \\ \frac{a}{1-a^2} & \frac{1}{1-a^2} + 1 \end{pmatrix}^{-1} \begin{pmatrix} \frac{1}{1-a^2} \\ \frac{a}{1-a^2} \end{pmatrix} = \dots = \begin{pmatrix} 2 \\ a \end{pmatrix} \frac{1}{4-a^2}\end{aligned}$$

The corresponding error is:

$$\begin{aligned}\mathbb{E}(X_n - \hat{X}_n)^2 &= R_x(0) - \rho_{xy}^* \tilde{a} = \frac{1}{1-a^2} - \begin{pmatrix} 1 & a \\ 1-a^2 & 1-a^2 \end{pmatrix} \tilde{a} \\ &= \dots = \frac{2}{4-a^2}\end{aligned}$$

(f) The Kalman filter equations are

$$\begin{aligned}\hat{X}_n &= a\hat{X}_{n-1} + P_n(Y_n - a\hat{X}_{n-1}) \\ P_n &= \frac{a^2 P_{n-1} + 1}{a^2 P_{n-1} + 2}, \quad n \geq 1\end{aligned}\tag{3.5}$$

subject to $\hat{X}_0 = 0$ and $P_0 = 1/(1-a^2)$.

(g) First note that $P_n \in [0, 1]$, since by optimality $P_n \leq E(Y_n - X_n)^2 = E\xi_n^2 = 1$. Let P_∞ be the unique nonnegative solution of

$$P_\infty = \frac{a^2 P_\infty + 1}{a^2 P_\infty + 2},\tag{3.6}$$

which is (the other solution is always negative)

$$P_\infty = \frac{a^2 - 2 + \sqrt{4 + a^4}}{2a^2}.$$

The sequence $D_n := |P_n - P_\infty|$ satisfies

$$\begin{aligned}D_n &= \left| -\frac{1}{a^2 P_{n-1} + 2} + \frac{1}{a^2 P_\infty + 2} \right| = \frac{a^2 D_{n-1}}{(a^2 P_{n-1} + 2)(a^2 P_\infty + 2)} \leq \\ &= \frac{a^2 D_{n-1}}{2(a^2(a^2 - 2 + \sqrt{4 + a^4})/(2a^2) + 2)} = \frac{a^2 D_{n-1}}{a^2 + 2 + \sqrt{4 + a^4}} \leq \frac{1}{2} D_{n-1}\end{aligned}$$

and thus $\lim_{n \rightarrow \infty} D_n = 0$.

The "steady state" filter is then

$$\begin{aligned}\hat{X}_n &= a\hat{X}_{n-1} + \frac{a^2 - 2 + \sqrt{4 + a^4}}{2a^2}(Y_n - a\hat{X}_{n-1}) = \\ &= a \left(1 - \frac{a^2 - 2 + \sqrt{4 + a^4}}{2a^2} \right) \hat{X}_{n-1} + \frac{a^2 - 2 + \sqrt{4 + a^4}}{2a^2} Y_n = \\ &= \underbrace{\frac{a^2 + 2 - \sqrt{4 + a^4}}{2a}}_{\equiv \gamma} \hat{X}_{n-1} + \frac{a^2 - 2 + \sqrt{4 + a^4}}{2a^2} Y_n.\end{aligned}$$

Note that this recursion is exactly the one which was obtained via Kolmogorov-Wiener approach in the appropriate setup.

- (h) The best estimate is obtained via optimal smoothing in (c); next is the filter, based on all the observations till n in (d). The Kalman filter in (f) is inferior to the latter filter for any fixed n , but is asymptotically equivalent to it as $n \rightarrow \infty$. The worst is of course the filter in (e) that takes into account only two observations. Note that for $a = 0$ (i.e. the signal X_n is an i.i.d. sequence (*white noise*), all the estimates attain the same error $P = 1/2$.
- (i) The error recursion for the Kalman filter becomes:

$$P_n = a^2 P_{n-1} - \frac{(a^2 P_{n-1})^2}{a^2 P_{n-1} + 1} = \frac{a^2 P_{n-1}}{a^2 P_{n-1} + 1}$$

This can be explicitly solved (define e.g. $Q_n = 1/P_n$ and obtain a linear recursion for Q_n) and verified that $\lim_{n \rightarrow \infty} P_n = 0$.

Problem 3.3

Recall that

$$X_n = \begin{cases} X_{n-1}, & \text{with prob. } p \\ -X_{n-1}, & \text{with prob. } 1-p \end{cases}$$

with $\mathbb{P}\{X_0 = \ell\} = \mathbb{P}\{X_0 = -\ell\} = 1/2$. Let $(\xi_n)_{n \geq 1}$ be an i.i.d. binary sequence of r.v. with

$$\mathbb{P}\{\xi_n = 1\} = 1 - \mathbb{P}\{\xi_n = 0\} = p.$$

Clearly $(X_n)_{n \geq 1}$ can be generated by:

$$X_n = (2\xi_n - 1)X_{n-1}, \quad \text{subject to } X_0$$

Rewrite this equation as:

$$X_n = (2\mathbb{E}\xi_n - 1)X_{n-1} + 2X_{n-1}(\xi_n - \mathbb{E}\xi_n) = (2p - 1)X_{n-1} + 2X_{n-1}(\xi_n - p)$$

Define $\eta_n = 2X_{n-1}(\xi_n - p)$, then:

$$\mathbb{E}\eta_n = 2\mathbb{E}X_{n-1}\mathbb{E}(\xi_n - p) = 0$$

and (say $n > m$)

$$\mathbb{E}\eta_n \eta_m = 4\mathbb{E}X_{n-1}X_{m-1}(\xi_m - p)\mathbb{E}(\xi_n - p) = 0$$

$$\mathbb{E}\eta_n^2 = 4\mathbb{E}X_{n-1}^2\mathbb{E}(\xi_n - p)^2 = 4\ell^2(1-p)p$$

Moreover for $k < n$, X_k and η_n are uncorrelated.

Introduce an auxiliary pair of processes (\tilde{X}, \tilde{Y}) , generated by

$$\begin{aligned} \tilde{X}_n &= (2p - 1)\tilde{X}_{n-1} + \tilde{\eta}_n, & \text{subject to } X_0 \\ \tilde{Y}_n &= \tilde{X}_n + \varepsilon_n. \end{aligned} \tag{3.7}$$

where $\tilde{\eta}$ is a white noise sequence with the same mean and variance as η .

The orthogonal projection $\hat{X}_n = \hat{E}(\tilde{X}_n | \tilde{Y}_1^n)$ is generated by the Kalman filter

$$\begin{aligned} \hat{X}_n &= (2p - 1)\hat{X}_{n-1} + P_n(\tilde{Y}_n - (2p - 1)\hat{X}_{n-1}), & n \geq 1 \\ P_n &= \frac{(2p - 1)^2 P_{n-1} + 4\ell^2 p(1-p)}{(2p - 1)^2 P_{n-1} + 4\ell^2 p(1-p) + 1} \end{aligned} \tag{3.8}$$

subject to $\hat{X}_0 = 0$ and $P_0 = \ell^2$. If these equation are applied to the original observations process Y , the obtained linear functional $\hat{X}_n(Y_1^n)$ can be considered

as an estimate for X_n . Does the obtained filter realizes the orthogonal projection $\widehat{E}(X_n|Y_1^n)$ for the original model?

Let $L(Y_1^n)$ denote any linear functional of $\{Y_1, \dots, Y_n\}$, then

$$E(X_n - L(Y_1^n))^2 = E(\widetilde{X}_n - L(\widetilde{Y}_1^n))^2 \geq E(\widetilde{X}_n - \widehat{X}_n(\widetilde{Y}_1^n))^2 = E(X_n - \widehat{X}_n(Y_1^n))^2, \quad (3.9)$$

where the equalities hold, since (X, Y) and $(\widetilde{X}, \widetilde{Y})$ have the same correlation structure by construction. The inequality (3.9) implies that $\widehat{X}_n(Y_1^n)$ is optimal and hence realizes the orthogonal projection.

Problem 3.4

Denote $\mu = E\eta_n$. Rewrite the eq. for Y_n as:

$$Y_n = \mu X_{n-1} + \xi_n + (\eta_n - \mu)X_{n-1}$$

Set $\widetilde{\xi}_n := \xi_n + (\eta_n - \mu)X_{n-1}$. Then:

$$\mathbb{E}\widetilde{\xi}_n = 0, \quad \mathbb{E}\widetilde{\xi}_n \widetilde{\xi}_k = \delta_{n-k}(\sigma_\xi^2 + \sigma_\eta^2 V_{n-1})$$

where $V_n = \mathbb{E}X_n^2$ satisfies ($n \geq 1$)

$$V_n = a^2 V_{n-1} + \sigma_\varepsilon^2, \quad \text{subject to } V_0 = 1$$

Moreover $\widetilde{\xi}_n$ is uncorrelated with X_m , $m < n$. Consider the model:

$$\begin{aligned} X_n &= aX_{n-1} + \varepsilon_n \\ Y_n &= \mu X_{n-1} + \widetilde{\xi}_n, \quad \text{s.t. } X_0 \end{aligned} \quad (3.10)$$

The optimal linear estimate is given by the Kalman filter

$$\begin{aligned} \widehat{X}_n &= a\widehat{X}_{n-1} + \frac{a\mu P_{n-1}}{\mu^2 P_{n-1} + \sigma_\xi^2 + \sigma_\eta^2 V_{n-1}} (Y_n - \mu\widehat{X}_{n-1}) \\ P_n &= a^2 P_{n-1} + \sigma_\varepsilon^2 - \frac{[a\mu P_{n-1}]^2}{\mu^2 P_{n-1} + \sigma_\xi^2 + \sigma_\eta^2 V_{n-1}} \end{aligned}$$

subject to $\widehat{X}_0 = 0$, $P_0 = 1$.

Problem 3.5

Simple Solution:

Define an augmented state vector $\vartheta_n \in \mathbb{R}^{(p+q) \times 1}$

$$\vartheta_n = \begin{bmatrix} \theta_n \\ \theta_{n-1} \\ \vdots \\ \theta_{n-p+1} \\ \varepsilon_n \\ \vdots \\ \varepsilon_{n-q+1} \end{bmatrix}$$

Introduce $A \in \mathbb{R}^{(p+q) \times (p+q)}$ and $B, C \in \mathbb{R}^{(p+q) \times 1}$:

$$A = \begin{bmatrix} -a_1 & -a_2 & \cdots & -a_p & b_1 & \cdots & b_q \\ 1 & 0 & 0 & \cdots & & & 0 \\ 0 & 1 & 0 & \cdots & & & 0 \\ \vdots & & & & & & \vdots \\ 0 & 0 & 0 & \cdots & & & 0 \\ 0 & 0 & & \cdots & 1 & & 0 \\ \vdots & & & & & & \vdots \\ 0 & 0 & & \cdots & & 1 & 0 \end{bmatrix}$$

and

$$B = \begin{bmatrix} 1 \\ 0 \\ \vdots \\ 0 \\ 1 \\ 0 \\ \vdots \\ 0 \end{bmatrix}, \quad C = \begin{bmatrix} 1 \\ 0 \\ \vdots \\ 0 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

Consider the vector difference equations ($n \geq p$):

$$\begin{aligned} \vartheta_n &= A\vartheta_{n-1} + B\varepsilon_n \\ \xi_n &= C^*\vartheta_n + v_n \end{aligned}$$

where ϑ_{p-1} is a vector of initial conditions (θ_0^{p-1} and ε_0^{q-1}). Clearly the first component of the vector ϑ_n coincides with θ_n for any $n \geq p$, i.e. $\vartheta_n(1) = \theta_n$. Note that $\mathbb{E}\vartheta_k\varepsilon_n = 0$, $k < n$ and hence the obtained model suits the Kalman filter setting: let $\hat{\vartheta}_n = \hat{\mathbb{E}}(\vartheta_n|\xi_0^n)$ and $\hat{\theta}_n = \hat{\mathbb{E}}(\theta_n|\xi_0^n)$, then ($n \geq p$):

$$\begin{aligned} \hat{\vartheta}_n &= A\hat{\vartheta}_{n-1} + \frac{(AP_{n-1}A^* + BB^*\sigma^2)C(\xi_n - C^*A\hat{\vartheta}_{n-1})}{C^*AP_{n-1}A^*C + C^*BB^*C\sigma^2 + \sigma_v^2} \\ P_n &= AP_{n-1}A^* + \sigma^2BB^* - \\ &\quad - \frac{(AP_{n-1}A^* + BB^*\sigma^2)CC^*(AP_{n-1}A^* + BB^*\sigma^2)}{C^*AP_{n-1}A^*C + C^*BB^*C\sigma^2 + \sigma_v^2} \\ \hat{\theta}_n &= C^*\hat{\vartheta}_n \end{aligned}$$

subject to $\hat{\vartheta}_{p-1} = 0$ and ${}^2P_{p-1} = I\sigma^2$.

Note that the estimates of $\{\theta_{n-1}, \dots, \theta_{n-p+1}\}$ and also of the driving noise $\{\varepsilon_n, \dots, \varepsilon_{n-q+1}\}$ are obtained as a byproduct.

Advanced Solution³: In the previous solution version to generate $\hat{\theta}_n$ one has to propagate $(p+q)$ -dimensional vector recursion. More delicate arguments lead to a

³ ${}^2\theta_0^{p-1}$ and ε_0^{q-1} are assumed to form a vector of i.i.d. components with zero mean and variance σ^2

filter of lower dimensions. Consider a sequence

$$\theta_n = -\sum_{k=1}^p a_k \theta_{n-k} + \sum_{k=0}^{p-1} b_k \varepsilon_{n-k} \quad (3.11)$$

where $(\varepsilon_n)_{n \geq 0}$ is standard white noise sequence. Note that the original model of the problem (i.e. $q \leq p$) is obtained by setting appropriate b_k 's to zero.

Below we derive a state space model of order p , which generates the same sequence.

Lemma 3.1. *Let η_n be a vector process generated by the recursion:*

$$\eta_n = A\eta_{n-1} + B\varepsilon_n, \quad n \geq 0 \quad (3.12)$$

where $(\varepsilon_n)_{n \geq 0}$ is an i.i.d. scalar standard Gaussian sequence and

$$A = \begin{pmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & & & \dots & \vdots \\ 0 & 0 & & \dots & 1 \\ -a_n & -a_{n-1} & & \dots & -a_1 \end{pmatrix}, \quad B = \begin{pmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{pmatrix}$$

$$\beta_1 = b_0$$

$$\beta_j = b_{j-1} - \sum_{\ell=1}^{j-1} a_{j-\ell} \beta_\ell, \quad j = 2, \dots, n$$

Then $\theta_n \equiv \eta_n(1)$.

Proof. Verify the equivalence between (3.11) and (3.12) as maps, i.e. show that both generate the same output $y(t), t = 0, 1, \dots$ for the same input $x(t), t = 0, 1, \dots$. Use the Z-transform system representation:

Starting from (3.12):

$$y_i(t) = y_{i+1}(t-1) + \beta_i x(t), \quad i = 1, \dots, n-1$$

$$y_n(t) = -\sum_{k=0}^{n-1} a_{n-k} y_{k+1}(t-1) + \beta_n x(t)$$

or

$$Y_i(z) = z^{-1} Y_{i+1}(z) + \beta_i X(z), \quad i = 1, \dots, n-1$$

$$Y_n(z) = -z^{-1} \sum_{k=0}^{n-1} a_{n-k} Y_{k+1}(z) + \beta_n X(z)$$

Then⁴ for $i = 1, \dots, n-1$

$$Y_{i+1}(z) = z[Y_i(z) - \beta_i X(z)] = \dots = z^i Y_1(z) - \sum_{j=1}^i z^{i-j+1} \beta_j X(z) \quad (3.13)$$

³This solution is for advanced reading.

⁴the convention $\sum_1^0 = 0$ is followed

On the other hand:

$$\begin{aligned}
Y_n(z) &= -z^{-1} \sum_{k=0}^{n-1} a_{n-k} Y_{k+1}(z) + \beta_n X(z) = \\
&= -z^{-1} \sum_{k=0}^{n-1} a_{n-k} z^k Y_1(z) + z^{-1} \sum_{k=0}^{n-1} a_{n-k} \sum_{j=1}^k z^{k-j+1} \beta_j X(z) + \\
&\quad + \beta_n X(z)
\end{aligned} \tag{3.14}$$

Equating the (3.13) with $i = n - 1$ and (3.14) we arrive at:

$$\begin{aligned}
z^n Y_1(z) + \sum_{k=0}^{n-1} a_{n-k} z^k Y_1(z) &= \sum_{j=1}^{n-1} z^{n-j+1} \beta_j X(z) + \\
+ \sum_{k=0}^{n-1} a_{n-k} \sum_{j=1}^k z^{k-j+1} \beta_j X(z) &+ z \beta_n X(z)
\end{aligned}$$

or ($a_0 := 1$)

$$\begin{aligned}
\sum_{k=0}^n a_{n-k} z^k Y_1(z) &= \sum_{j=1}^{n-1} z^{n-j+1} \beta_j X(z) + \\
+ \sum_{k=0}^{n-1} a_{n-k} \sum_{j=1}^k z^{k-j+1} \beta_j X(z) &+ z \beta_n X(z)
\end{aligned}$$

and

$$\begin{aligned}
Y_1(z) \sum_{k=0}^n a_k z^{-k} &= \sum_{j=1}^{n-1} z^{-j+1} \beta_j X(z) + \\
+ \sum_{k=1}^{n-1} a_{n-k} \sum_{j=n-k+1}^n z^{-j+1} \beta_{j-n+k} X(z) &+ z^{-(n-1)} \beta_n X(z)
\end{aligned} \tag{3.15}$$

Equating the right hand side of (3.15) to $X(z)P_{n-1}(z)$ and comparing powers of z we obtain the desired result:

$$\begin{aligned}
z^0 &: \beta_1 = b_0 \\
z^{-1} &: \beta_2 + a_1 \beta_1 = b_1 \\
&\vdots \\
z^{-(n-1)} &: \sum_{k=1}^{n-1} a_{n-k} \beta_k + \beta_n = b_{n-1}
\end{aligned}$$

□

Let us demonstrate the latter approach:

Example

Let ξ_n be a stationary random process with zero mean and the spectrum density:

$$f(\lambda) = \left| \frac{1 + e^{-j\lambda}}{1 + 1/2e^{-j\lambda} + 1/2e^{-2j\lambda}} \right|^2$$

Find the optimal linear extrapolation estimate of ξ_t on the basis of $\{\xi_0, \dots, \xi_s\}$, $m(t, s) = \widehat{\mathbb{E}}(\xi_t | \xi_0^s)$.

Find the state space representation for ξ_n . Here $b_0 = 1, b_1 = 1$ and $a_0 = 1, a_1 = 1/2, a_2 = 1/2$ and thus $\beta_1 = b_0 = 1$ and $\beta_2 = b_1 - 1/2 \cdot 1 = 1/2$. Let $(\eta_1(t), \eta_2(t))$ be generated by

$$\begin{aligned}\eta_1(t) &= \eta_2(t-1) + \varepsilon(t) \\ \eta_2(t) &= -1/2\eta_1(t-1) - 1/2\eta_2(t-1) + 1/2\varepsilon(t)\end{aligned}$$

where $\varepsilon(t)$ is a standard i.i.d. Gaussian sequence.

Set $\xi_t = \eta_1(t)$ and $\theta_t = \eta_2(t)$. Then ξ_t has the spectral density $f(\lambda)$ and:

$$\begin{aligned}\xi_t &= \theta_{t-1} + \varepsilon(t) \\ \theta_t &= -1/2\theta_{t-1} - 1/2\xi_{t-1} + 1/2\varepsilon(t)\end{aligned}\tag{3.16}$$

And thus ($t > s$)

$$\begin{aligned}m(t, s) &= \mu(t-1, s) \\ \mu(t, s) &= -1/2\mu(t-1, s) - 1/2m(t-1, s)\end{aligned}$$

subject to $m(s, s) = \xi_s$ and $\mu(s, s) = \mu(s) = \widehat{\mathbb{E}}(\theta_s | \xi_0^s)$.

The filtering estimate $\mu(s)$ satisfies ($k \leq s$):

$$\begin{aligned}\mu_k &= -1/2\mu_{k-1} - 1/2\xi_{k-1} + \frac{1/2 - 1/2P_{k-1}}{P_{k-1} + 1}(\xi_k - \mu_{k-1}) \\ P_k &= 1/4P_{k-1} + 1/4 - \frac{(1/2 - 1/2P_{k-1})^2}{P_{k-1} + 1} = \frac{P_{k-1}}{P_{k-1} + 1}\end{aligned}\tag{3.17}$$

The initial conditions for this filter can be recovered due to stationarity assumptions. Let $d_{11} = \mathbb{E}\theta_t^2$, $d_{12} = \mathbb{E}\xi_t\theta_t$ and $d_{22} = \mathbb{E}\xi_t^2$. From (3.16):

$$\begin{aligned}d_{22} &= d_{11} + 1 \\ d_{11} &= 1/4d_{11} + 1/4d_{22} + 1/4 + 1/2d_{12} \\ d_{12} &= -1/2d_{11} - 1/2d_{12} + 1/2\end{aligned}$$

so that:

$$d_{11} = 1, \quad d_{12} = 0, \quad d_{22} = 2$$

and the initial condition for the filter (3.17):

$$\mu_0 = 0, \quad P(0) = 1$$

Problem 3.6

The Riccati equation of the Kalman filter is transformed by Matrix Inversion Lemma into:

$$\begin{aligned}P_{n+1} &= aP_n a^* + bb^* - aP_n A^* (AP_n A^* + BB^*)^{-1} AP_n a^* = \\ &= bb^* + a \{ P_n - P_n A^* (AP_n A^* + BB^*)^{-1} AP_n \} a^* = \\ &= bb^* + a\Gamma_n^{-1} a^*\end{aligned}$$

where

$$\Gamma_n = P_n^{-1} + A^*(BB^*)^{-1}A = J_n + A^*(BB^*)^{-1}A$$

$$\begin{aligned} J_{n+1} &:= P_{n+1}^{-1} = \left\{ bb^* + (a^{-*}\Gamma_n a^{-1})^{-1} \right\}^{-1} = \\ &= F_n - F_n b (I + b^* F_n b)^{-1} b^* F_n \end{aligned}$$

where $F_n := a^{-*}\Gamma_n a^{-1}$. Summarizing all the equations, J_n can be propagated by:

$$\begin{aligned} J_{n+1} &= F_n - F_n b (I + b^* F_n b)^{-1} b^* F_n \\ F_n &= a^{-*} (J_n + A^* (BB^*)^{-1} A) a^{-1} \end{aligned}$$

The validity of the Matrix Inversion Lemma is verified directly:

$$\begin{aligned} AA^{-1} &= (B^{-1} + CD^{-1}C^*)(B - BC(D + C^*BC)^{-1}C^*B) = \\ &= I + CD^{-1}C^*B - C(D + C^*BC)^{-1}C^*B - \\ &\quad - CD^{-1}C^*BC(D + C^*BC)^{-1}C^*B = \\ &= I + CD^{-1}C^*B - C \{I + D^{-1}C^*BC\} (D + C^*BC)^{-1}C^*B = \\ &= I + CD^{-1}C^*B - CD^{-1}C^*B \equiv I \end{aligned}$$

Problem 3.7

Clearly x_t are the orthogonal projections of a standard random vector x on $\{y_1, \dots, y_t\}$, where

$$y_{t+1} = a_{t+1}x + \sqrt{\alpha}\varepsilon_{t+1}$$

with ε_t being standard white noise, independent of x . So x_k is the orthogonal projection of x on $y = Ax + \sqrt{\alpha}\varepsilon$, where ε is a standard random vector.

Then

$$Q = \mathbb{E}(xy^*)(\mathbb{E}(yy^*))^{-1} = A^*(\alpha I + AA^*)^{-1} = (\alpha I + A^*A)^{-1}A^*$$

since

$$A^*(\alpha I + AA^*) = (\alpha I + A^*A)A^*$$

The second statement follows from the fact that $\gamma_k = \mathbb{E}(x - x_k)(x - x_k)^*$

$$\begin{aligned} \gamma_k &= \mathbb{E}(xx^*) - \mathbb{E}(xy^*)\mathbb{E}^{-1}(yy^*)\mathbb{E}(yx^*) = I - A^*(\alpha I + AA^*)^{-1}A = \\ &= (I + A^*A/\alpha)^{-1} = (I\alpha + A^*A)^{-1}\alpha \end{aligned}$$